An Efficient Carry Select Adder with Less Delay and Reduced Area Application

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Abstract- Design of area, high speed and power-efficient data path logic systems forms the largest areas of research in VLSI system design. The addition speed is limited by the time necessary to transmit a carry through the adder. Carry Select Adder (CSLA) is one of the fastest adders used in several dataprocessing processors to perform fast arithmetic purpose. From the configuration of the CSLA, it is clear that there is scope for decreasing the area and delay in the CSLA. This work uses a simple and an efficient gate-level modification which drastically reduces the area and delay of the CSLA. Based on this modification 16, 32, 64 and 128-bit square-root Carry Select Adder (SORT CSLA) architectures have been improved and compared with the regular SQRT CSLA architecture. The proposed design has compact area and delay to a great extent when compared with the regular SQRT CSLA. This work estimates the performance of the planned designs with the regular designs in terms of delay, area and synthesis are implemented in Xilinx FPGA. The results analysis shows that the proposed CSLA structure is better than the regular SQRT CSLA.

Keywords—(ASIC), area-efficient, CSLA, low delay.

I. INTRODUCTION

Reduced area and high speed data path logic systems are the major areas of research in VLSI system design. High-speed addition and multiplication has always been a fundamental necessity of high-performance processors and systems. In digital adders, the speed of addition is partial by the time necessary to propagate a carry through the adder. The sum for each bit position in an elementary adder is generated serially only after the previous bit position has been summed and a carry propagated into the next position. There are several types of adder designs available (RCA, CLAA, CSA, CSA) which have its own advantages and disadvantages. The main speed limitation in any adder is in the production of carries and many authors considered the addition problem. To solve the carry propagation delay CSLA is developed which drastically decreases the area and delay to a

great extent. The CSLA is used in many computational systems design to moderate the problem of carry propagation delay by separately generating multiple carries and then select a carry to generate the sum. It uses independent ripple carry adders (for Cin=0 and Cin=1) to generate the resultant sum. However, the Regular CSLA is not area and speed efficient because it uses multiple pairs of Ripple Carry Adders (RCA) to generate partial sum and carry by considering carry input. The final sum and carry are selected by the multiplexers (mux). Due to the use of two independent RCA the area will enlarge which leads an increase in delay. To overcome the above problem, the basic idea of the planned work is to use n-bit binary to excess-1 code converters (BEC) to improve the speed of addition. This logic can be replaced in RCA for Cin=1 to further improves the speed and thus decreases the delay. Using Binary to Excess-1 Converter (BEC) instead of RCA in the regular CSLA will achieve lower area, delay which speeds up the addition operation. The major advantage of this BEC logic comes from the lesser number of logic gates than the Full Adder (FA) structure because the number of gates used will be decreased.

This work in brief is structured as follows. Section II deals with the delay and area evaluation methodology of the basic adder blocks and its corresponding delay and area values. Section III deals with the structure and function of BEC logic and its corresponding function table and logic equations. Section IV presents the architecture of the Regular CSLA of 128-bits. This SQRT CSLA has been developed using ripple carry adders and multiplexers. The architecture of the Modified SQRT CSLA is presented in Sections V. In section VI implementation methodologies and corresponding design tools are explained and finally the paper is concluded in section VIII.

II. BASIC ADDER BLOCK

The adder block with a Ripple carry adder, BEC and Mux is explained in this section. In this we calculate and explain the delay & area using the theoretical approach and explain how the delay and area effect the total implementation. The AND, OR, and Inverter (AOI) implementation of an XOR gate is shown in Fig. 1. The delay and area evaluation methodology considers all gates to be made up of AND, OR, and Inverter, each having delay equal to 1 unit and area equal to 1 unit. We then add up the number of gates in the longest path of a logic block that contributes to the maximum delay. The area evaluation is done by counting the total number of AOI gates necessary for each logic block. Based on this approach, the blocks of 2:1 mux, Half Adder (HA), and FA are evaluated in Table I.



Fig 1: Delay and area evalution of xor

TABLE 1: DELAY AND AREA EVALUATION OF CSLA

Design	Delay	Area
XOR	3	5
2:1 MUX	3	4
Half Adder	3	6
Full Adder	6	13



Fig 2: 8-bit BEC with 16:8 mux

Fig 2 shows the basic 8-bit addition procedure which includes 8-bit data, a 8-bit BEC logic and 16:8 mux. The addition operation is performed for Cin=0 and for Cin=1.For Cin=0 the addition is performed using ripple carry adder and for Cin=1 the operation is performed using 8-bit BEC (replacing the RCA for Cin=1). The resultant is selected based on Carry in signal from the earlier group. The total delay depends on mux delay and Cin signal from earlier group.

III. BINARY TO EXCESS-1 CONVERTER (BEC)

The basic work is to use Binary to Excess-1 Converter (BEC) in the regular CSLA to get lower area and improved speed of operation. This logic is replaced in RCA with Cin=1. This logic can be implemented for different bits which are used in the modified design. The major advantage of this BEC logic comes from the fact that it uses lesser number of logic gates than the n-bit Full Adder (FA) structure. As stated above the main idea of this work is to use BEC instead of the RCA with Cin=1 in order to decrease the area and increase the speed of operation in the regular CSLA to obtain modified CSLA. To replace the n-bit RCA, an n+1 bit BEC logic is required. The structure and the function table of a 8-bit BEC are shown in Figure 2 respectively.

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Figure 3: 6-binary to excess-1 converter

TABLE 2: FUNCTION TABLE OF THE 6-BIT BEC

B[5:0]	X[5:0]
000000	000001
000001	000010
000010	000011
	000000
111111	000000

The Boolean expressions for the 6-bit BEC logic are expressed below

X0 = ~B0

 $X1 = B0^{A}B1$

 $X2 = B2^{(B0 \& B1)}$

X3 = B3^ (B0 & B1 & B2)

X4 = B4^ (B0 & B1 & B2 & B3)

X5 = B5^ (B0 & B1 & B2 & B3 & B4).

 $X6 = B6^{(B0 \& B1 \& B2 \& B3 \& B4 \& B5).$

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X7 = B^{(B0 \& B1 \& B2 \& B3 \& B4 \& B5 \& B6).
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Figure 4: Architecture of Regular 128-bit SQRT CSLA

IV. ARCHITECTURE OF REGULAR 128-BIT SQRT CSLA

A 16-bit carry select adder can be designed in two different sizes namely uniform block size and variable block size. Similarly a 32, 64 and 128-bit can also be developed in two modes of different block sizes. Ripple-carry adders are the simplest and most compact full adders, but their performance is limited by a carry that must spread from the least-significant bit to the most-significant bit. The various 16, 32, 64 and 128-bit CSLA can also be developed by using ripple carry adders. The speed of a carry-select adder can be enhanced upto 40% to 90%, by performing the additions in parallel, and reducing the maximum carry delay. Fig 4 shows the Regular structure of 128-bit SQRT CSLA. It includes many ripple carry adders of variable sizes which are separated into groups.

Group 0 contains 2-bit RCA which contains only one ripple carry adder which adds the input bits and the input carry and results to sum [1:0] and the carry out. The carry out of the Group 0 which acts as the selection input to mux which is in group 1, selects the result from the consequent RCA (Cin=0) or RCA (Cin=1). Similarly the remain groups will be selected depending on the Cout from the previous groups.

In Regular CSLA, there is only one RCA to achieve the addition of the least significant bits [1:0]. The remaining bits (other than LSBs), the addition is performed by using two RCAs corresponding to the one assuming a carry-in of 0, the other a carry-in of 1 within a group. In a group, there are two RCAs that receive the same data inputs but different Cin. The upper adder has a carry-in of 0, the lower adder a carry-in of 1. The actual Cin from the previous sector selects one of the two RCAs. That is, as shown in the Fig.4, if the carry-in is 0, the sum and carry-out of the upper RCA is selected, and if the carry-in is 1, the sum and carry-out of the lower RCA is selected. For this Regular CSLA architecture, the execution code, for the Full Adders and Multiplexers of different sizes (6:3, 8:4, 10:5 up to 24:11) were designed initially. The regular 64-bit, 128-bit CSLA were implemented by calling the ripple carry adders and all multiplexers.

V. ARCHITECTURE OF MODIFIED 128-BIT SQRT CSLA

This architecture is alike to regular 128-bit SQRT CSLA, the only change is that, we change RCA with Cin=1 among the two available RCAs in a group with a BEC. This BEC has a feature that it can perform the alike operation as that of the replaced RCA with Cin=1. Fig 5 shows the Modified block diagram of 128-bit SQRT CSLA. The number of bits necessary for BEC logic is 1 bit more than the RCA bits. The modified block diagram is also separated into various groups of variable sizes of bits with each group having the ripple carry adders, BEC and corresponding mux . As shown in the Fig.5, Group 0 include one RCA only which is having input of lower significant bit and carry in bit and produces result of sum[1:0] and carry out which is acting as mux selection line for the next group, also the procedure continues for higher groups but they includes BEC logic instead of RCA with Cin=1. Based on the consideration of delay values, the arrival time of selection input C1 of 8:3 mux is before than the sum of RCA and BEC. For remain groups the selection input arrival is later than the RCA and BEC. Thus, the sum1 and c1 (output from mux) are depending on mux and results computed by RCA and BEC respectively. The sum 2 depends on c1 and mux. For the remain parts the arrival time of mux selection input is always larger than the arrival time of data inputs from the BEC's. Thus, the delay of the remaining MUX depends on the arrival time of mux selection input and the mux delay. In this Modified CSLA architecture, the implementation code for Full Adder and Multiplexers of 6:3, 8:4, and 10:5 up to 24:11 were designed. The design code for the BEC was designed by using NOT, XOR and AND gates. Then 2, 3, 4, 5 up to 11-bit ripple carry adder was designed.



Figure 5: Architecture of Modified 128-bit SQRT CSLA

TABLE 3: COMPARISON VALUES OF 16 BIT, 32 BIT,64 BIT AND128 BIT CSLA

$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	Sr. No	Adders		Delay (ns)	Area (No. of LUTs)
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	1 16	16 hit	Regular	18.42	43
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		10 - bh	Modified	16.56	47
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	2 32 – bit	32 hit	Regular	25.53	92
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		32 - 600	Modified	21.41	101
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	3 64 – bit	61 bit	Regular	42.17	189
4 128 - bit Regular 56.37 422 Modified 40.56 439		04 - 011	Modified	25.95	206
4 $128 - 01$ Modified 40.56 430	4	128 bit	Regular	56.37	422
Modified 40:30 439		128 - 01	Modified	40.56	439

VI. RESULTS

The implemented design in this work has been simulated using Verilog-HDL. The adders (of various size 16, 32, 64 and 128) are designed and simulated with Modelsim. All the V files (Regular and Modified) are also simulated in Modelsim and corresponding results are compared. After simulation the different size codes are synthesized using Xilinx ISE 9.1i. The simulated V files are imported into the synthesized tool and resultant values of delay and area are noted. The synthesized reports include area and delay values for different sized adders. The similar design flow is followed for both the regular and modified SQRT CSLA of different sizes.

Table 3 shows the comparison of regular and modified CSLA of various bits which includes Delay and area comparisons. From the table it is clear that the delay decreases for 16-bit modified method when compared with regular method.

Similarly the table also shows the comparison for the various 32, 64, and 128 bits.

The relative values of areas shows that the number of LUT will be more for modified the 16, 32 and 64. This value reduces gradually for 128 bits. For 256 bits the value almost equal to regular method which will decreases more for still higher order bits. Thus the modified method reduces the delay and also area to a great extent.

VII. CONCLUSION

An efficient approach is planned in this paper to decrease the area and delay of SQRT CSLA architecture. The reduction in the number of gates is obtained by simply replacing the RCA with BEC in the structure. The compared results shows that the modified SQRT CSLA has a slightly bigger area for lower order bits which further reduces for higher order bits. The delay is reduced to a great extent with the modified SQRT CSLA. Thus the results shows that using modified method the area and delay will reduce thus leads to good alternative for adder implementation for many processors. The modified CSLA architecture is therefore low area and high speed approaches for VLSI hardware implementation.

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